# High Efficiency, Synchronous 5A Buck Charger for 1 cell Li-ion Battery with NVDC Power Path Management

# 1 DESCRIPTION

The SC89890H is a 1.5MHz highly integrated switch-mode buck charger for 1 cell Li-ion battery applications and NVDC system power path management, which separate the system load and charge current, also the system can power up with deep depletion battery. System can get the power from VBUS, VBAT or both. It supports 3.9-13.5V input voltage, up to 5A charging current and provide battery charge management functions including trickle charge, constant current charge, constant voltage charge, charge termination, auto recharge and charging status indication.

The SC89890H supports flexible charge current option, and the user can program the current freely through external resistor for different applications. With the charger management function, the IC can be used to charge 1 cell Li-ion battery.

The SC89890H supports USB OTG with up to 2.4A output with PFM/PWM mode. Meanwhile, the SC89890H supports USB Input and High Voltage fast charge Adapters, auto detect USB BC1.2, SDP, CDP, DCP.

The SC89890H supports input current and voltage limit, input under voltage and over voltage protections, internal cycle by cycle current limit, battery short circuit protection, and output over voltage protection. It also offers charging safety timer and over temperature protection to ensure safety under different abnormal conditions.

The SC89890H integrated all MOSFETs, current sensing, loop compensation and I2C interface.

The SC89890H is available in QFN(24)-4\*4 package.

# 2 FEATURES

- Integrated Synchronous Buck Charger
- Integrated NVDC Power Path Management
- Charging Management (Trickle Charge / Constant Current Charge / Constant Voltage Charge / Charge Termination)
- Integrated I2C interface
- I2C Programmable Constant Charge Current, ±5%
   @1.5A-5A accuracy
- I2C Programmable Constant Voltage, ±0.5% accuracy
- I2C Programmable Charge Safety Timer
- Support OTG discharging function and I2C Programmable Output Voltage: 3.9V~5.4V with up 2.4A current
- Support Shipping mode, Low battery leakage current
- Charge Status Indication
- Integrated ADC for System Monitor
- Resistance Compensation (IRCOMP) from Charger Output to Cell Terminal and ICO Support
- NTC for Battery Protection (support JEITA standard)
- Input Under Voltage and Over Voltage Protection
- Internal Cycle by Cycle Over Current Protection
- OTG OCP/OVP Protection
- Battery Over Voltage and Short Protection
- Battery Discharging Over Current and under voltage Protection
- Thermal Regulation and Shutdown
- QFN (24)-4\*4 footprint

# **3 APPLICATIONS**

- Smart Phones
- Portable Internet Devices and Accessory

# **4 DEVICE INFORMATION**

Part Number	Package	Dimension
SC89890HQDLR	QFN(24)-4*4	4x4

# 5 Typical Application Circuit



# 6 Terminal Configurations and Functions



QFN(24) 4x4 (TOP View, SC89890H)

	I/O		DESCRIPTION
SC89890H	NAME		
1	VBUS	I	Power supply pin. Place a 1uF ceramic capacitor from VBUS to GND close to the IC
2	DP	ю	Positive line of the USB data line pair. DP/DM based USB host/charging port detection. The detection includes data contact detection (DCD), primary and secondary detection in BC1.2, and Adjustable high voltage adapter. The pin can be configured as output driver by DP_DAC register bits after USB Detection finished or during OTG mode.
3	DM	Ю	Negative line of the USB data line pair. DP/DM based USB host/charging port detection. The detection includes data contact detection (DCD), primary and secondary detection in BC1.2, and Adjustable high voltage adapter. The pin can be configured as output driver by DM_DAC register bits after USB Detection finished or during OTG mode.
4	STAT	0	Open-drain charge status output. Connect the STAT pin to a logic rail via 10-kΩ resistor. The STAT pin indicates charger status. Collect a current limit resister and a LED from a rail to this pin. Charge in progress: LOW Charge complete , charger in SLEEP mode and charger disable: HIGH Charge suspend (fault response): 1-Hz, 50% duty cycle Pulses .
5	SCL	I	I2C interface clock. Connect SCL to the logic rail through a 10-k $\Omega$ resistor.
6	SDA	IO	I2C interface data. Connect SDA to the logic rail through a 10-k $\Omega$ resistor.
7	/INT	0	Open-drain interrupt Output. Connect the INT to a logic rail through $10-k\Omega$ resistor. The INT pin sends an active low, 256-µs pulse to host to report charger device status and fault.
8	OTG	I	Active high enable pin during boost mode. The boost mode is activated when OTG_CONFIG =1 and OTG pin is high
9	/CE	I	Active low Charge Enable pin. Battery charging is enabled when CHG_CONFIG = 1 and /CE pin = Low. /CE pin must be pulled High or Low.
10	ILIM	I	Input current limit Input. ILIM pin sets the maximum input current and can be used to monitor input current . ILIM pin sets the maximum input current limit by regulating the ILIM voltage at 0.8 V. A resistor is connected from ILIM pin to ground to set the maximum limit as $IIN_{MAX} = K_{ILIM}/R_{ILIM}$ . The actual input current limit is the lower limit set by ILIM pin (when

			<ul> <li>EN_ILIM bit is high) or IIINLIM register bits. Input current limit of less than 500 mA is not support on ILIM pin.</li> <li>ILIM pin can also be used to monitor input current when the voltage is below 0.8V. The input current is proportional to the voltage on ILIM pin and can be calculated by IIN = (KILIM x VILIM) / (RILIM x 0.8)</li> <li>The ILIM pin function can be disabled when EN_ILIM bit is 0.</li> </ul>
11	NTC	ю	Connect to the Negative Temperature Coefficient (NTC) thermistor inside the battery cells to sense the battery cells temperature for protection.
12	/QON	I	BATFET enable/reset control input. When BATFET is in ship mode, a logic low of $t_{SHIPMODE}$ duration turns on BATFET to exit shipping mode. When VBUS is not plugged–in, a logic low of $t_{QON_RST}$ (minimum 8 s) duration resets SYS (system power) by turning BATFET off for $t_{BATFET_RST}$ (minimum 250 ms) and then re-enable BATFET to provide full system power reset. The pin contains an internal pull-up to maintain default high logic.
13,14	VBAT	0	Battery connection point to the positive terminal of the battery pack. Connect a 10uF ceramic capacitor close to the VBAT pin.
15,16	VSYS	0	Converter output connection point. Connect a 20 $\mu\text{F}$ capacitor close to the VSYS pin.
17,18	PGND	I	Power ground pin.
19,20	SW	0	Switching node output. Connected to output inductor. Connect the 47nF bootstrap capacitor from SW to BTST.
21	BTST	ю	PWM high side driver positive supply. Internally, the BTST pin is connected to the cathode of the boost-strap diode. Connect the 47nF bootstrap capacitor from SW to BTST.
22	VCC	0	HSFET and LSFET driver and internal supply output. Internally, VCC is connected to the anode of the boost-strap diode. Connect a $4.7-\mu F$ (10-V rating) ceramic capacitor from VCC to GND. The capacitor should be placed close to the IC.
23	PMID	0	Connected to the drain of the reverse blocking MOSFET (RBFET) and the drain of HSFET. Put 10 $\mu$ F ceramic capacitor on PMID to GND.
24	DSEL	0	Active high DP/DM multiplexer selection control. Connect a 47-nF (6V rating) ceramic capacitor from DSEL to analog GND. The pin is normally low. During input source type detection, the pin drives high to indicate the device DP/DM detection is in progress and needs to take control of DP, DM signals. When detection is completed, the pin keeps high when DCP, HVDCP is detected. The pin returns to low when other input source type is detected.

# 7 Specification

## 7.1 Absolute Maximum Rating

Over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

		Min.	Max.	Unit
	VBUS	-0.3	22	V
	PMID	-0.3	22	V
Voltage <sup>(2)</sup>	BTST	-0.3	22	V
Vollage	SW <sup>(3)</sup>	-2(10ns)	16	V
	BTST to SW	-0.3	6	V
	DP,DM,VCC,NTC,/CE,VBAT,VSYS,SDA,SCL,/INT,/QON,ILIM, STAT ,DSEL	-0.3	6	V
TJ	Operating junction temperature	-40	150	°C
T <sub>stg</sub>	Storage temperature	-65	150	°C

(1) Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device.

(2) All voltages are with respect to network ground terminal.

(3) VBUS, PMID, BTST, SW should be tied together to test the SW abs voltage.

## 7.2 Thermal Information

THERMAL RESISTA	THERMAL RESISTANCE <sup>(1)</sup>		Unit
θ <sub>JA</sub>	Junction to ambient thermal resistance	34	°C/W
θ <sub>JC</sub>	Junction to case resistance	25	°C/W

(1) Measured on JESD51-7, 4-layer PCB.

## 7.3 ESD Ratings

			Min.	Max.	Unit
V <sub>ESD</sub> <sup>(1)</sup>	Human-body Model (HBM) (2)	All pins	-2	+2	kV
	Charged-device Model (CDM) (3)		-750	750	V

(1) Electrostatic discharge (ESD) to measure device sensitivity and immunity to damage caused by assembly line electrostatic discharges into the device.

(2) Level listed above is the passing level per ANSI, ESDA, and JEDEC JS-001. JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(3) Level listed above is the passing level per EIA-JEDEC JESD22-C101. JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

## 7.4 Recommended Operation Conditions

		MIN	TYP	МАХ	UNIT
V <sub>BUS</sub>	VBUS voltage range	3.9		13.5	V
VBAT	VBAT voltage range		4.2	4.848	V
l <sub>iN</sub>	Input current limit			3.25	А

lcc	Constant current charge current (SW Output Current)			5	A
lava	Discharging current continuous	6			А
I <sub>DIS</sub>	Discharging current (1s)	10			
L	Inductance		1		μH
TA	Operating ambient temperature	-40		85	°C
TJ	Operating junction temperature	-40		125	°C

# 8 Function Block Diagram



# 9 Electrical Characteristics

 $T_{A}\text{=}~25~^\circ\text{C}~$  and  $V_{AC\_UVLO}\text{<}~V_{BUS}\text{<}~V_{VAC\_OVP}\text{, unless otherwise noted.}$ 

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
SUPPLY VOL	TAGE					
V <sub>BUS</sub>	Operating input V <sub>BUS</sub> voltage		3.9		13.5	V
		Rising edge		3.3		V
V <sub>VBUS_UVLO</sub>	V <sub>BUS</sub> for active I2C, no battery	Hysteresis		300		mV
		Falling edge	20	100	210	mV
V <sub>SLEEP</sub>	V <sub>BUS</sub> -V <sub>BAT</sub> threshold	Rising edge	120	220	370	mV
	V <sub>BUS</sub> Over Voltage threshold	Rising edge	13.5	14.2	14.9	V
V <sub>VAC_OVP</sub>		Hysteresis		300		mV
.,		Rising edge	2.25			V
V <sub>VBAT_UVLO</sub>	BAT for active I2C, no VBUS	Hysteresis		240		mV
	Detterre Derdetter three held	Rising edge	2.5	2.6	2.75	V
V <sub>VBAT_DPL</sub>	Battery Depletion threshold	Hysteresis		200		mV
	Bad adapter detection	Falling edge	3.6	3.7	3.8	V
VVBUSMIN	threshold	Hysteresis		200		mV
BADSRC	Bad adapter detection sink current from $V_{\text{BUS}}$ to GND			30		mA
I <sub>BAT</sub>	Battery discharge current in Buck mode	$V_{BAT}$ = 4.5 V, HIZ mode, no $V_{BUS}$ , BATFET_DIS Enable ,ADC Disable		15	25	uA
		$V_{BAT}$ = 4.5 V, HIZ mode, no $V_{BUS}$ , BATFET_DIS Disable ,ADC Disable		22	35	uA
	Input supply current in buck	V <sub>BUS</sub> =5V, HIZ mode and BATFET _DIS Disable, no battery			40	uA
I <sub>VBUS</sub> _HIZ	mode when HIZ mode is enabled	V <sub>BUS</sub> =12V,HIZ mode and BATFET _DIS Disable, no battery			60	uA
		V <sub>BUS</sub> > V <sub>VBUS_UVLO</sub> , V <sub>BUS</sub> >V <sub>BAT</sub> , Converter not switching		1.5	3	mA
I <sub>VBUS</sub>	Input supply current in buck mode	$\label{eq:VBUS} \begin{array}{ll} V_{BUS} > V_{VBUS\_UVLO}, & V_{BUS} > V_{BAT}, \\ Converter & switching, & V_{BAT} = 3.8V, \\ I_{SYS} = 0A, & disable & charger \end{array}$		3		mA
I <sub>BOOST</sub>	Battery discharge current in boost mode	V <sub>BAT</sub> =4.2V,boost mode, I <sub>VBUS</sub> =0A, converter switching			3	mA
POWER PAT	H					
V <sub>SYS</sub>	Typical system regulation voltage	I <sub>SYS</sub> =0A,V <sub>BAT</sub> <v<sub>SYSMIN,I<sub>SYS</sub>=0A, BATFET Disable</v<sub>		V <sub>SYSMIN</sub> +0 .18		V
V <sub>SYS_MIN</sub>	Minimum system regulation voltage	V <sub>VBAT</sub> < V <sub>SYS_MIN</sub> = 3.5V, BATFET Disabled	3.55	3.68		V
	vollage	Range	3		3.7	V
V <sub>SYS_MAX</sub>	Maximum DC system voltage output	I <sub>SYS</sub> =0A,V <sub>BAT&gt;</sub> V <sub>SYSMIN</sub> ,I <sub>SYS</sub> =0A, BATFET Disable, Vv <sub>BAT</sub> <=4.4V	4.4	4.45	4.51	v

R <sub>DSON_Q1</sub>	Reverse blocking MOSFET on resistance	V <sub>cc</sub> =5V		35	46	mΩ
R <sub>DSON_Q2</sub>	High side switching MOSFET on resistance	V <sub>cc</sub> =5V		38	46	mΩ
$R_{DSON_Q3}$	Low side switching MOSFET on resistance	V <sub>cc</sub> =5V		25	36	mΩ
R <sub>DSON_Q4</sub>	$V_{\text{SYS}}$ to $V_{\text{BAT}}$ MOSFET on resistance	V <sub>VBAT</sub> =4.2V		17	25	mΩ
V <sub>FWD</sub>	Supplement mode Q4 forward voltage			50		mV
CHARGER MA	NAGEMENT					
VBATREG_RANGE	Regulation Charge Voltage		3.84		4.848	V
V <sub>BATREG_STEP</sub>	Charge Voltage Step			16		mV
		V <sub>BATREG</sub> =4.208V	4.18	4.208	4.23	V
$V_{\text{BATREG}_{ACC}}$	Charge Voltage accuracy	V <sub>BATREG</sub> =4.384V	4.36	4.384	4.4	V
		V <sub>BATREG</sub> =4.432V	4.4	4.432	4.45	V
I <sub>CC_RANGE</sub>	Constant charging current range		0		5.04	А
I <sub>CC_STEP</sub>	Constant charging current step			60		mA
	Constant charging current accuracy	I <sub>CC</sub> =240mA, V <sub>BAT</sub> = 3.8V	216	240	264	mA
		$I_{CC} = 720 \text{mA}, V_{BAT} = 3.8 \text{V}$	684	720	756	mA
		I <sub>CC</sub> =2040mA, V <sub>BAT</sub> = 3.8V	1938	2040	2142	mA
		$I_{CC} = 3000 \text{mA}$ , $V_{BAT} = 3.8 \text{V}$	2850	3000	3150	mA
	Trickle charge to CC Charge	Rising edge	2.9	3	3.1	V
V <sub>TC</sub>	battery voltage threshold	Hysteresis		200		mV
I <sub>TC_RANGE</sub>	Trickle charge current range		60		960	mA
I <sub>TC_STEP</sub>	Trickle charge current step			60		mA
		I <sub>TC</sub> =120mA	90	120	150	mA
I <sub>TC_ACC</sub>	Trickle charge current accuracy	I <sub>TC</sub> =420mA	390	420	450	mA
ITERM_RANGE	Termination current range		30		930	mA
I <sub>TERM_STEP</sub>	Termination current step			60		mA

		I <sub>TERM</sub> =120mA	90	120	150	mA
I <sub>TERM_ACC</sub>	Termination current accuracy	I <sub>TERM</sub> =240mA	216	240	264	mA
		I <sub>TERM</sub> =420mA	390	420	450	mA
		Falling edge	1.85	2	2.15	V
V <sub>BAT_SHORT</sub>	Battery short voltage	Hysteresis		200		mV
I <sub>BAT_SHORT</sub>	Battery short charge current	V <sub>BAT</sub> =1V	35	50	65	mA
	Recharge threshold below	V <sub>BAT</sub> falling edge,100mV	80	100	140	mV
V <sub>RECHG</sub>	V <sub>BAT_REG</sub>	V <sub>BAT</sub> falling edge, 200mV	170	200	250	mV
ISYSLOAD	System discharge load current	V <sub>SYS</sub> =4.2V		30		mA
t <sub>TERM_DGL</sub>	Deglitch time for charge termination			250		ms
t <sub>RECH_DGL</sub>	Deglitch time for recharge			20		ms
IBAT_OCP_DGL	System over-current(10A) deglitch time to turn off Q4			100		us
t <sub>SYSOVP_DGL</sub>	System over-voltage deglitch time to turn off DCDC			1		us
t <sub>BATOVP_DGL</sub>	Battery over-voltage deglitch time to disable charger			1		us
INPUT VOL	TAGE AND CURRENT REC	GULATION				1
VINDPM_RANGE	Input voltage regulation limit range		3.9		15.3	V
VINDPM_STEP	Input voltage regulation limit step			100		mV
N/	Input voltage regulation limit	V <sub>INDPM</sub> =4.8V	4.6	4.8	4.944	V
VINDPM_ACC	accuracy	V <sub>INDPM</sub> =8.2V	7.9	8.4	8.5	V
N/		V <sub>INDPM_OS</sub> =400mV		400		mV
VINDPM_OS	VINDPM=VBUS - VINDPM_OS	V <sub>INDPM_OS</sub> =600 mV		600		mV
IINDPM_RANGE	Input current regulation limit range		100		3250	mA
IINDPM_STEP	Input current regulation limit step			50		mA
		V <sub>BUS</sub> =5V, I <sub>INDPM</sub> =500 mA	410		500	mA
INDPM_ACC	Input current regulation limit accuracy	V <sub>BUS</sub> =5V,I <sub>INDPM</sub> =900 mA	750		900	mA
		V <sub>BUS</sub> =5V,I <sub>INDPM</sub> =1500 mA				1

		V <sub>BUS</sub> =5V,I <sub>INDPM</sub> =2400 mA	2200		2400	mA
I <sub>IN_START</sub>	Input current limit during system start-up sequence			200		mA
K <sub>ILM</sub>	I <sub>INMAX</sub> =K <sub>ILM</sub> /R <sub>ILM</sub>	Input current regulation by ILIM pin = 1.5 A	310	350	385	AxΩ
PROTECTION	N					1
		Rising	102	104	105	%
$V_{VBAT_OVP}$	Battery over voltage threshold	Hysteresis		2		%
I <sub>BAT_OCP</sub>	Battery discharge over current threshold	100µs deglitch	10			А
PWM	1					
f <sub>sw</sub>	PWM switching frequency	$V_{BUS}$ = 9V, $V_{BAT}$ =4V, $I_{CC}$ = 3A	1250	1500	1680	KHz
D <sub>MAX</sub>	Maximum PWM duty cycle(Buck)			97%		
JEITA (BUCK	(MODE)					
V <sub>COLD</sub>	NTC cold temp (0°C) threshold,	Rising	72.4	73.3	74.2	%
V COLD	as percentage of VCC	falling	71.5	72	72.5	%
		5°C Rising	70.25	70.75	71.25	%
		5°C falling	68.7	69.2	69.7	%
	NTC cool temp threshold, as percentage of VCC	10°C Rising	67.75	68.25	68.75	%
V <sub>COOL</sub>		10°C falling	66.45	66.95	67.45	%
V COOL		15°C Rising	64.75	65.25	65.75	%
		15°C falling	63.7	64.2	64.7	%
		20°C Rising	61.75	62.25	62.75	%
		20°C falling	60.7	61.2	61.7	%
		40°C Falling	47.74	48.25	48.75	%
		40°C Rising	48.8	49.3	49.8	%
		45°C Falling	44.25	44.75	45.25	%
	NTC warm temp threshold, as	45°C Rising	45.3	45.8	46.3	%
V <sub>WARM</sub>	percentage of VCC	50°C Falling	40.2	40.7	41.2	%
		50°C Rising	41.3	41.8	42.3	%
		55°C Falling	37.2	37.7	38.2	%
		55°C Rising	38.5	39	39.5	%
	NTC hot temp (60°C) threshold,	Falling	33.7	34.2	34.7	%
V <sub>HOT</sub>	as percentage of VCC	Rising	34.8	35.3	35.8	%
	Charging current during JEITA	20%		20		%
RATIO_COOL	COOL as percentage of I <sub>CC</sub>	50%		50		%
	Charging current during JEITA	20%		20		%
RATIO_WARM	WARM as percentage of I <sub>CC</sub>	50%		50		%

	V <sub>BAT</sub> regulation voltage in JEITA	100mV		100		mV
V <sub>DELTA_WARM</sub>	WARM equal to $V_{\text{BATREG}}$ – $V_{\text{DELTA}_{\text{WARM}}}$	200mV		200		mV
NTC (BOOST N	IODE)					
V	NTC cold temp threshold, as	Rising	79.5	80	80.5	%
V <sub>BCOLD</sub>	percentage of VCC	falling	78.5	79	79.5	%
Vauaa	NTC hot temp threshold, as	Falling	29.7	31.2	32	%
V BHO1	percentage of VCC	Rising	33.9	34.4	34.9	%
BOOST MODE	OPERATION					
$V_{\text{OTG}\_\text{REG}\_\text{RANGE}}$	Boost mode regulation voltage range		3.9		5.4	V
Votg_reg_step	Boost mode regulation voltage step			100		mV
	Boost mode regulation voltage	V <sub>OTG_REG</sub> =5V, V <sub>BAT</sub> =3.8V, I <sub>BUS</sub> =0A	4.85	5	5.15	V
Votg_reg_acc	accuracy	$V_{OTG\_REG}$ =5.2V, $V_{BAT}$ =3.8V, $I_{BUS}$ =0A	5.04	5.2	5.36	V
		V <sub>VBAT</sub> falling,2.8V	2.7	2.8	2.9	V
Vvbatlow_otg	Battery voltage exiting boost	Hysteresis		200		mV
	mode	V <sub>VBAT</sub> falling,2.5V	2.4	2.5	2.6	V
		Hysteresis		300		mV
I <sub>OTG_RANGE</sub>	OTG mode output current limit range		0.5		2.45	А
I <sub>OTG_ACC</sub>	OTG mode output current limit accuracy	I <sub>OTG</sub> =1.2A	1.2	1.4	1.6	А
V <sub>OTG_OVP</sub>	OTG overvoltage threshold	Rising		5.8		V
VCC LDO	l					
V <sub>VCC</sub>	V <sub>cc</sub> LDO output voltage	V <sub>BUS</sub> =9V, I <sub>VCC</sub> =20mA		5		V
I <sub>vcc</sub>	V <sub>CC</sub> current limiter	VBUS=5V, $V_{VCC}$ = 3.8V, Charger disable	50			mA
LOGIC IO						1
V <sub>ILO</sub>	Input low threshold				0.4	V
V <sub>IHO</sub>	Input high threshold		0.9			V
/QON TIMING	I					
t <sub>shipmode</sub>	/QON low time to turn on BATFET and exit ship mode		0.9		1.3	s
t <sub>QON_RST</sub>	/QON low time to reset BATFET		8		12	s
t <sub>BATFET_RST</sub>	BATFET off time during full system reset		250		400	ms
t <sub>BATFET_DLY</sub>	Enter ship mode delay		10		15	s
DIGITAL CLOC	K AND WATCHDOG TIMER	1				1
t <sub>WDT</sub>	Watchdog timer			40		s
f <sub>SCL</sub>	SCL Clock frequency				400	KHz

t <sub>TC</sub>	Safety timer for Trickle charge			4		hours
t <sub>CC/CV</sub>	Safety timer for CC and CV			12		
VBUS Power	r up					•
t <sub>VAC_OVP</sub>	V <sub>BUS</sub> OVP reaction time			100		ns
t <sub>BADSRC</sub>	Bad adapter detection duration			30		ms
THERMAL R	EGULATION and SHUTDOWN	I				
T <sub>REG</sub>	The second second strengthere is a second second	Range	60		120	°C
	Thermal regulation temperature	Step		20		°C
т	Thermal shutdown temperature			150		°C
T <sub>SD</sub>	Thermal shutdown hysteresis			30		°C
DP/DM Deteo	ction					•
$V_{0\text{P6}_{VSRC}}$	DP/DM voltage source		0.5	0.6	0.7	V
$V_{1P2\_VSRC}$	DP/DM voltage source		1.1	1.2	1.3	V
$V_{\text{2P0}_{VSRC}}$	DP/DM voltage source		1.9	2	2.1	V
$V_{\text{2P7}_\text{VSRC}}$	DP/DM voltage source		2.6	2.7	2.8	V
V <sub>3P3_VSRC</sub>	DP/DM voltage source		3.2	3.3	3.4	V
$V_{0P325\_VTH}$	DP/DM Input comparator threhsold		0.25		0.4	V
$V_{1P0\_VTH}$	DP/DM Input comparator threhsold		0.9	7	1.1	V
$V_{1P35_VTH}$	DP/DM Input comparator threhsold		1.25		1.45	V
$V_{\text{2P2\_VTH}}$	DP/DM Input comparator threhsold		2.1		2.3	V
$V_{3\text{P0}\_\text{VTH}}$	DP/DM Input comparator threhsold		2.9		3.1	V
		0.6V output	250			μA
DP/DM_SRC	BC1.2 DP/DM source capability	3.3V output	250			μA
R <sub>DP/DM_PD</sub>	DP/DM pull down resistor			19.53		КΩ
DP/DM_SINK	BC1.2 DP/DM sink current			50		μA
Pump Curre	nt Pulse					
tpumpx_stop	Current pulse control stop pulse		430		570	ms
t <sub>PUMPX_ON1</sub>	Current pulse control long on pulse		240		360	ms
t <sub>PUMPX_ON2</sub>	Current pulse control short on pulse		70		130	ms
t <sub>PUMPX_OFF</sub>	Current pulse control off pulse		70		130	ms

tpumpx_dly	Current pulse control stop start delay		80		225	ms
DSEL						
V <sub>OL</sub>	Output low threshold level	C <sub>DSEL</sub> =47nF			0.4	V
V <sub>OH</sub>	Output high threshold level	C <sub>DSEL</sub> =47nF	4.5			V
IR COMP						
Resistance		Range	0		140	mΩ
		Step		20		mΩ
V <sub>CLAMP</sub>		Range	0		224	mV
		Step		32		mV

# **10 Feature Description**

### 10.1 Power-On-Reset(POR)

The SC89890H powers internal bias circuits from the higher voltage of VBUS and VBAT. When VBUS rises above  $V_{VBUS\_UVLO}$  or VBAT rises above  $V_{VBAT\_UVLO}$ , the sleep comparator, battery depletion comparator and BATFET driver are active. I2C interface is ready for communication and all the registers are reset to default value. The host can access all the registers after POR.

# 10.2 Device Power Up from Battery without Input Source

If only battery is present and the voltage is above depletion threshold , the BATFET turns on and connects battery to system. The VCC LDO stays off to minimize the quiescent current. The low  $R_{DSON}$  of BATFET and the low quiescent current on BAT minimize the conduction loss and maximize the battery run time.

The device always monitors the discharge current through BATFET (Supplement Mode). When the system is overloaded or shorted ( $I_{BAT} > I_{BAT_OCP}$ ), the device turns off BATFET immediately and set BATFET\_DIS bit to indicate BATFET is disabled until the input source plugs in again or one of the methods described in BATFET Enable (Exit Shipping Mode) is applied to re-enable BATFET.

# 10.3 Power Up from Input Source

When an input source is plugged in, the device checks the input source voltage to turn on VCC LDO and all the bias circuits. It detects and sets the input current limit before the buck converter is started. The power up sequence from input source is as listed:

- 1. Power up VCC LDO
- 2. Poor Source Qualification

3. Input Source Type Detection is based on DP/DM to set default input current limit (IINDPM) register or input source type

4. Input Voltage Limit Threshold Setting (VINDPM threshold)

5. Converter Power-up

### 10.3.1 Power Up VCC LDO Regulation

The VCC LDO supplies internal bias circuits as well as the HSFET and LSFET gate drive. The VCC also provides bias rail to NTC external resistors. The pull-up rail of STAT can

be connected to VCC as well. The VCC is enabled when all the below conditions are valid:

- V<sub>BUS</sub> above V<sub>VBUSMIN</sub> , above V<sub>BAT</sub> + V<sub>SLEEP</sub> in buck mode
- VBUS below VBAT + VSLEEP in boost mode
- Above conditions are satisfied during 220ms delay

If any one of the above conditions is not valid, the device is in high impedance mode (HIZ) with VCC LDO off. The device draws less than  $I_{\text{VBUS}\_\text{HIZ}}$  from VBUS during HIZ state. The battery powers up the system when the device is in HIZ mode.

By setting EN\_HIZ bit to 1 with adapter, the device enters high impedance state (HIZ). In HIZ mode, the system is powered from battery even with good adapter present. The device is in the low input quiescent current state with Q1 RBFET, VCC LDO and the bias circuits off.

### 10.3.2 Poor Source Qualification

After VCC LDO powers up, the device confirms the current capability of the input source. The input source must meet both of the following requirements in order to start the buck converter:

- VBUS voltage below V<sub>VAC\_OVP</sub>
- VBUS voltage above V<sub>VBUSMIN</sub> when pulling I<sub>BADSRC</sub> (typical 30 mA)

Once the input source passes all the conditions above, the status register bit VBUS\_GD is set high and the /INT pin is pulsed to signal to the host. If the device fails the poor source detection, it repeats poor source qualification every 2 seconds.

### 10.3.3 Input Source Type Detection

After the VBUS\_GD bit is set and VCC LDO is powered, the device runs input source detection through DP/DM. The SC89890H follows the USB Battery Charging Specification 1.2 (BC1.2) to detect input source (SDP/CDP/DCP) and non-standard adapter through USB DP/DM lines.

After input source type detection is completed, an INT pulse is asserted to the host. in addition, the following registers and pin are changed:

1. Input Current Limit (IINDPM) register is changed to set current limit

2. PG\_STAT bit is set

3. VBUS\_STAT bit is updated to indicate USB or other input source

The host can over-write IINDPM register to change the input current limit if needed. The charger input current is always limited by the IINDPM register. The charger input current is always limited by the lower of IINLIM register or ILIM pin at all-time regardless of Input Current Optimizer (ICO) is enable or disabled.

# When AUTO\_DPDM\_EN is disabled, the Input Source Type Detection is bypassed.

The SC89890H contains a DP/DM based input source detection to set the input current limit at VBUS plug-in. The DP/DM detection includes standard USB BC1.2 and non-standard adapter. When input source is plugged in, the device starts standard USB BC1.2 detection. The USB BC1.2 is capable to identify Standard Downstream Port (SDP), Charging Downstream Port (CDP) and Dedicated Charging Port (DCP). When the Data Contact Detection (DCD) timer expires, the nonstandard adapter detection is applied to set the input current limit. The non-standard detection is used to distinguish vendor specific adapters (Apple and Samsung) based on their unique dividers on the DP/D– pins. If an adapter is detected as DCP, the input current limit is set at 3.25 A. If an adapter is detected as unknown, the input current limit is set at 0.5 A.



Figure1 USB DP/DM Detection

DP/DM DETECTION	INPUT CURRENT LIMIT
USB SDP	500mA
USB CDP	1.5A
USB DCP	3.25A
Divider 1	2.1A
Divider 2	2A
Divider 3	1A
Divider 4	2.4A
Unknown Adapter	500mA

Table1 Input current limit setting from DP/DM Detection

#### 10.3.4 Input Voltage Limit Threshold Setting

The device supports wide range of input voltage limit (3.9 V - 14 V) for high voltage charging and provides two methods to set Input Voltage Limit (VINDPM) threshold to facilitate autonomous detection.

#### 1. Absolute VINDPM (FORCE\_VINDPM=1)

By setting FORCE\_VINDPM bit to 1, the VINDPM threshold setting algorithm is disabled. Register VINDPM is writable and allows host to set the absolute threshold of VINDPM function.

2. Relative VINDPM based on VINDPM\_OS registers (FORCE\_VINDPM=0)

When FORCE\_VINDPM bit is 0 (default), the VINDPM threshold setting algorithm is enabled. The VINDPM register is read only and the charger controls the register by using VINDPM threshold setting algorithm (VBUS-VINDPM\_OS).

After Input Voltage Limit Threshold is set, an INT pulse is generated to signal to the host

#### 10.3.5 Converter Power-Up

After the input current limit is set, the converter is enabled and the HSFET and LSFET start switching. If battery charging is disabled, BATFET turns off. Otherwise, BATFET stays on to charge the battery.

The SC89890H provides soft-start when system rail is ramped up. When the system rail is below 2.2 V, the input current is limited to is to the lower of 200 mA or IINDPM

register setting. After the system rises above 2.2 V, the device limits input current to the value set by IINDPM register.

As a battery charger, the device deploys a highly efficient 1.5 MHz step-down switching regulator. The fixed frequency oscillator keeps tight control of the switching frequency under all conditions of input voltage, battery voltage, charge current and temperature, simplifying output filter design.

The SC89890H switches to PFM control at light load or when battery is below minimum system voltage setting or charging is disabled.

### 10.4 Boost Mode Operation From Battery

The SC89890H supports boost converter operation to deliver power from the battery to other portable devices through USB port. The boost mode output current rating meets the USB On-The-Go 500 mA output requirement. The maximum output current is up to 2.4 A. The boost operation can be enabled if the conditions are valid:

- 1. VBAT above V<sub>VBATLOW\_OTG</sub>
- 2. VBUS less than VBAT + VSLEEP
- 3. Set OTG\_CONFIG and OTG Pin high
- 4. Battery is not in BCOLD and BHOT.
- 5. Above conditions are satisfied during 30ms delay.

During boost mode, the status register VBUS\_STAT bits is set to 111, the VBUS output is 5 V and the output current can reach up to 2.4 A, selected through I2C ( $I_{BOOST}$  bit). The boost output is maintained when BAT is above  $V_{VBATLOW_OTG}$  threshold.

In boost mode, the device employs a 500 KHz or 1.5 MHz (selectable using BOOST\_FREQ bit) step-up switching regulator based on system requirements. To avoid frequency change during boost mode operations, write to boost frequency configuration bit (BOOST\_FREQ) is ignored when OTG\_CONFIG is set.

When OTG is enabled, the device starts up with PFM and later transits to PWM to minimize the overshoot. The PFM\_OTG\_DIS bit can be used to prevent PFM operation in either buck or boost configuration.

### 10.5 Host Mode and Default Mode

The SC89890H is a host controlled charger, but it can operate in default mode without host management. In default mode, the device can be used as an autonomous charger with no host or while host is in sleep mode. When the charger is in default mode, WATCHDOG\_FAULT bit is HIGH. When the charger is in host mode, WATCHDOG\_FAULT bit is LOW.

After power-on-reset, the device starts in default mode with watchdog timer expired, or default mode. All the registers are in the default settings. in default mode, the device keeps charging the battery with default 10-hour fast charging safety timer. At the end of the 10-hour, the charging is stopped and the buck converter continues to operate to supply system load.

Writing reg transitions the charger from default mode to host mode. All the device parameters can be programmed by the host. To keep the device in host mode, the host has to reset the watchdog timer by writing 1 to WD\_RST bit before the watchdog timer expires (WATCHDOG\_FAULT bit is set), or disable watchdog timer by setting WATCHDOG bits = 00.

When the watchdog timer expires (WATCHDOG\_FAULT bit = 1), the device returns to default mode and all registers are reset to default values except IINDPM, VINDPM, VINDPM\_OS, BATFET\_RST\_EN, BATFET\_DLY, and BATFET\_DIS bits.





### 10.6 ICO

The device provides innovative Input Current Optimizer (ICO) to identify maximum power point without overload the input source. The algorithm automatically identify maximum input current limit of power source without entering VINDPM to avoid input source overload.

This feature is enabled by default (ICO\_EN=1) and can be disabled by setting ICO\_EN bit to 0. After DCP type input source is detected based on the procedures previously The algorithm runs automatically when ICO\_EN bit is set. The algorithm can also be forced to execute by setting FORCE\_ICO bit regardless of input source type detected.

The actual input current limit used by the Dynamic Power Management is reported in IDPM\_LIM register while Input Current Optimizer is enabled (ICO\_EN = 1) or set by IINLIM register when the algorithm is disabled (ICO\_EN = 0). In addition, the current limit is clamped by ILIM pin unless EN\_ILIM bit is 0 to disable ILIM pin function.

# 10.7 NVDC Power Path Management

The SC89890H accommodates a wide range of input sources from USB, wall adapter, to car charger. The device provides automatic power path selection to supply the system (VSYS) from input source (VBUS), battery (VBAT), or both.

### 10.7.1 Battery Charging Management

The SC89890H charges 1-cell Li-lon battery with up to 5A charge current for high capacity tablet battery. The low Rdson BATFET improves charging efficiency and minimize the voltage drop during discharging.

### 10.7.1.1 Autonomous Charging Cycle

With battery charging is enabled (CHG\_CONFIG bit = 1 and /CE pin is LOW), the device autonomously completes a charging cycle without host involvement. The host can always control the charging operations and optimize the charging parameters by writing to the corresponding registers through I2C.

Charging Parameters	Default Value
Charging Voltage	4.208V
CC Current	2.040A
TC Current	120mA
Termination Current	270mA
Battery Temperature Profile	JEITA
Safety Timer	TC:4hours, CC/CV:12hours

#### **Table2 Charging Parameter Default Setting**

A new charge cycle starts when the following conditions are valid:

- Converter starts
- Battery charging is enabled (CHG\_CONFIG bit =1, Icc is not 0A and /CE is low)
- No NTC COLD or HOT fault
- No safety timer fault
- BATFET is not forced to turn off (BATFET\_DIS bit=0)

The charger device automatically terminates the charging cycle when the charging current is below termination threshold, battery voltage is above recharge threshold, and device not is in DPM mode or thermal regulation. When a fully charged battery is discharged below recharge threshold (selectable through VRECHG bit), the device automatically starts a new charging cycle. After the charge is done, toggle CE pin or CHG\_CONFIG bit can initiate a new charging cycle. Adapter removal and re plug will also start a new charging cycle.

The STAT output indicates the charging status: charging (LOW), charging complete or charge disable (HIGH) or charging fault (Blinking). The STAT output can be disabled by setting STAT\_DIS=1. in addition, the status register (CHRG\_STAT) indicates the different charging phases: 00-charging disable, 01-precharge, 10-fast charge (constant current) and constant voltage mode, 11-charging done. Once a charging cycle is completed, an INT is asserted to notify the host.

STAT status	IC working status					
Low	Normal charging (TC/CC/CV/Recharge)					
High	End of charging (EOC, top off timer maybe running) , charge disable, sleep mode, Boost Mode					
1Hz Blinking	Charge suspend(VBUS OVP, NTC COLD/HOT, Safety timer out, VBAT OVP). Boost Mode suspend(NTC/COLD/HOT)					

Table3 STAT Pin status

### 10.7.1.2 Battery Charging Profile

The SC89890H charges the battery in five phases: battery short, TC,CC,CV, and top-off trickle charging (optional). At the beginning of a charging cycle, the device checks the battery voltage and regulates current and voltage accordingly.

V <sub>VBAT</sub>	Charging current	Default value	CHRG_STAT
< V <sub>BAT_SHORT</sub>	IBAT_SHORT	50mA	01
V <sub>BAT_SHORT</sub> to V <sub>TC</sub>	I <sub>TC</sub>	120mA	01
> V <sub>TC</sub>	I <sub>cc</sub>	2.040A	10

#### **Table4 Charging Current Setting**

If the charger device is in DPM regulation or thermal regulation during charging, the actual charging current will be less than the programmed value. in this case, termination is temporarily disabled and the charging safety timer is doubled.



Figure3 Battery Charging Profile

#### 10.7.1.3 End of Charge

The SC89890H terminates a charge cycle when the battery voltage is above recharge threshold, and the current is below termination current. After the charging cycle is completed, the BATFET turns off. The converter keeps running to power the system, and BATFET can turn on again to engage Supplement Mode.

When termination occurs, the status register CHRG\_STAT is set to 11, and an INT pulse is asserted to the host. Termination is temporarily disabled when the charger device is in input current, voltage or thermal regulation. Termination can be disabled by writing 0 to EN\_TERM bit prior to charge termination.

#### 10.7.1.4 NTC in Buck mode

The SC89890H monitors the battery cells' temperature through NTC pin. It monitors the NTC voltage. Once it detects the temperature is below 0°C or higher than 60°C, the IC will stop charging. Below shows the NTC operation summary.

V <sub>NTC</sub>	Temperature	Operation							
V <sub>NTC</sub> > V <sub>COLD</sub>	T < 0°C	Stop charging							
$V_{COLD} > V_{NTC} > V_{COOL}$	0°C < T < 10°C	0/0.5/0.2/1 CC current							
V <sub>COOL</sub> >V <sub>NTC</sub> > V <sub>WARM</sub>	10°C < T < 45°C	Normal charging							
V <sub>WARM</sub> > V <sub>NTC</sub> > V <sub>HOT</sub>	45°C < T < 60°C	CV/CV-50m/CV- 100mV/CV-200mV							
		0/0.5/0.2/1 CC current							

Table5 NTC function



**Figure4 NTC function** 

#### 10.7.1.5 NTC in Boost mode

For battery protection during boost mode, the SC89890H monitors the battery temperature to be within the  $V_{BCOLD}$  to  $V_{BHOT}$  thresholds. When temperature is outside of the temperature thresholds, the boost mode is suspended. In additional, VBUS\_STAT bits are set to 000 and NTC\_FAULT is reported. Once temperature returns within thresholds, the boost mode is recovered and NTC\_FAULT is cleared.

#### 10.7.1.6 Safety Timer

The SC89890H has built-in safety timer to prevent extended charging cycle due to abnormal battery conditions. The safety timer is 4 hours when the battery is below  $V_{TC}$  threshold and 12 hours when the battery is higher than  $V_{TC}$  threshold.

The user can program CC/CV charge safety timer through I2C (CHG\_TIMER bits). When safety timer expires, the fault register CHRG\_FAULT bits are set to 11 and an INT is asserted to the host. The safety timer feature can be disabled through I2C by setting EN\_TIMER bit.

During input voltage, current, JEITA cool/warm or thermal regulation, the safety timer will double as the setting value. The timer double function can be disable by writing 0 to TMR2X\_EN bit.

During the fault(BAT\_FAULT,NTC\_FAULT), timer is suspended. Once the fault goes away, timer resumes. If user stops the current charging cycle, and start again, timer gets reset (toggle /CE pin or CHR\_CONFIG bit).

#### 10.7.2 Resistance Compensation

For high current charging system, resistance between charger output and battery cell terminal such as board routing, connector, MOSFETs and sense resistor can force the charging process to move from constant current to constant voltage too early and increase charge time. To speed up the charging cycle, the device provides resistance compensation (IRCOMP) feature which can extend the constant current charge time to delivery maximum power to battery.

The device allows the host to compensate for the resistance by increasing the voltage regulation set point based on actual charge current and the resistance as shown below. For safe operation, the host should set the maximum allowed regulation voltage register ( $V_{CLAMP}$ ) and the minimum resistance compensation (BAT\_COMP).

 $V_{REG_{ACTUAL}} = V_{REG} + min(I_{CHRG_{ACTUAL}} x BAT_{COMP}, V_{CLAMP})$ 

### 10.7.3 Battery Monitor

The device includes a battery monitor to provide measurements of VBUS voltage, VBUS current, battery voltage, system voltage, ntc ratio and charging current based on the device modes of operation. The measurements are reported in Battery Monitor Registers (REG0E-REG12). The battery monitor can be configured as two conversion modes by using CONV\_RATE bit: one-shot conversion (default) and 1 second continuous conversion.

For one-shot conversion (CONV\_RATE = 0), the CONV\_START bit can be set to start the conversion. During the conversion, the CONV\_START is set and it is cleared by the device when conversion is completed. The conversion result is ready after  $t_{CONV}$  (maximum 1 second).

For continuous conversion (CONV\_RATE = 1), the CONV\_RATE bit can be set to initiate the conversion. During active conversion, the CONV\_START is set to indicate conversion is in progress. The battery monitor provides conversion result every 1 second automatically. The battery monitor exits continuous conversion mode when CONV\_RATE is cleared.

When battery monitor is active, the VCC power is enabled and can increase device quiescent current. In battery only mode, the battery monitor is only active when V(BAT) > SYS\_MIN setting.

### 10.7.4 Input Current Limit on ILIM

For safe operation, the device has an additional hardware pin on ILIM to limit maximum input current on ILIM pin. The input maximum current is set by a resistor from ILIM pin to ground as:

$$I_{\text{INMAX}} = \frac{K_{\text{ILIM}}}{R_{\text{ILIM}}}$$

The actual input current limit is the lower value between ILIM setting and register setting (IINLIM). For example, if the register setting is 111111 for 3.25 A, and ILIM has a 260- $\Omega$  resistor (KILIM = 390 max.) to ground for 1.5 A, the input current limit is 1.5 A. ILIM pin can be used to set the input current limit rather than the register settings when EN\_ILIM bit is set. The device regulates ILIM pin at 0.8 V. If ILIM voltage exceeds 0.8 V, the device enters input current regulation.

The ILIM pin can also be used to monitor input current when EN\_ILIM is enabled. The voltage on ILIM pin is proportional to the input current. ILIM pin can be used to monitor the input current following Equation:

$$I_{\rm IN} = \frac{K_{\rm ILIM} \times V_{\rm ILIM}}{R_{\rm ILIM} \times 0.8 \, \rm V}$$

For example, if ILIM pin is set with  $260-\Omega$  resistor, and the ILIM voltage is 0.4 V, the actual input current 0.615 A - 0.75 A (based on KILM specified). If ILIM pin is open, the input current is limited to zero since ILIM voltage floats above 0.8 V. If ILIM pin is short, the input current limit is set by the register.

The ILIM pin function can be disabled by setting EN\_ILIM bit to 0. When the pin is disabled, both input current limit function and monitoring function are not available.

### 10.7.5 Current Pulse Protocol

The device provides the control to generate the VBUS current pulse protocol to communicate with adjustable high voltage adapter in order to signal adapter to increase or decrease output voltage. To enable the interface, the EN\_PUMPX bit must be set. Then the host can select the increase/decrease voltage pulse by setting one of the PUMPX\_UP or PUMPX\_DN bit (but not both) to start the VBUS current pulse sequence. During the current pulse sequence, the PUMPX\_UP and PUMPX\_DN bits are set to indicate pulse sequence is in progress and the device pulses the input current limit between current limit set forth by IINLIM or IDPM\_LIM register and the 100mA current limit (IINDPM100\_ACC). When the pulse sequence is completed, the input current limit is returned to value set by IINLIM or IDPM\_LIM register and the PUMPX\_UP or PUMPX\_DN bit is cleared. In addition, the EN\_PUMPX can be cleared during the current pulse sequence to terminate the sequence and force charger to return to input current limit as set forth by the IINLIM or IDPM\_LIM register immediately. When EN\_PUMPX bit is low, write to PUMPX\_UP and PUMPX\_DN bit would be ignored and have no effect on VBUS current limit.

If the load in VBUS (including battery charging) less than 100mA, the current pulse may not work.

#### 10.7.6 DP/DM DAC

The device provides independent controlled voltage output drivers on DP and DM pins to interface or emulate nonstandard adapters when input source is plugged-in or OTG mode is enabled. The DP/DM drivers are disabled in high impedance mode (HiZ) by default or when DP or DM Control bits are set to be HIZ.

The drivers are enabled and controlled independently with predefined voltage threshold(HiZ, Short also supported)

The host is recommended to change DP and DM settings after input source type detection when VBUS\_STAT/PG\_STAT bits are updated.

When OTG mode is enabled, the drivers can be enabled to provide electrical signature on DP/DM to emulate USB nonstandard adapters

#### 10.7.7 NVDC Architecture

The SC89890H deploys Narrow VDC architecture (NVDC) with BATFET separating system from battery. The minimum system voltage is set by  $V_{SYS\_MIN}$  bits. Even with a fully depleted battery, the system is regulated above the minimum system voltage.

When the battery is below minimum system voltage setting, the BATFET operates in linear mode (LDO mode), and the system is typically 180 mV above the minimum system voltage setting. As the battery voltage rises above the minimum system voltage, BATFET is fully on and the voltage difference between the system and battery is the  $V_{DS}$  of BATFET.

When the battery charging is disabled and above minimum system voltage setting or charging is terminated, the system is always regulated at typically 50mV above battery voltage. The status register VSYS\_STAT bit goes high when the system is in minimum system voltage regulation.



Figure5 System Voltage vs Battery Voltage

#### 10.7.7.1 Dynamic Power Management

To meet maximum current limit in USB spec and avoid over loading the adapter, the device features Dynamic Power management (DPM), which continuously monitors the input current and input voltage. When input source is over-loaded, either the current exceeds the input current limit (IIDPM) or the voltage falls below the input voltage limit (VINDPM). The device then reduces the charge current until the input current falls below the input current limit and the input voltage rises above the input voltage limit.

When the charge current is reduced to zero, but the input source is still overloaded, the system voltage starts to drop. Once the system voltage falls below the battery voltage, the device automatically enters the supplement mode where the BATFET turns on and battery starts discharging so that the system is supported from both the input source and battery.

During DPM mode, the status register bits VINDPM\_STAT (VINDPM) or IINDPM\_STAT (IINDPM) goes high. Below figure shows the DPM response with 9-V/1.2-A adapter, 3.2-V battery, 2.8-A charge current and 3.5-V minimum system voltage setting.



#### 10.7.7.2 Supplement Mode

When the system voltage falls below the battery voltage, the BATFET turns on and the BATFET gate is regulated the gate drive of BATFET so that the minimum BATFET  $V_{DS}$  stays at 30 mV when the current is low. This prevents oscillation from entering and exiting the supplement mode.

As the discharge current increases, the BATFET gate is regulated with a higher voltage to reduce  $R_{DSON}$  until the BATFET is in full conduction. At this point onwards, the BATFET  $V_{DS}$  linearly increases with discharge current. BATFET turns off to exit supplement mode when the battery is below battery depletion threshold.



Figure7 BATFET V-I Curve

### 10.8 Shipping Mode and /QON Pin

### 10.8.1 BATFET Disable Mode(Shipping Mode)

To extend battery life and minimize power when system is powered off during system idle, shipping, or storage, the device can turn off BATFET so that the system voltage is zero to minimize the battery leakage current. When the host set BATFET\_DIS bit, the charger can turn off BATFET immediately or delay by  $t_{BATFET_DLY}$  as configured by BATFET\_DLY bit.

### 10.8.2 BATFET Enable(Exit Shipping Mode)

When the BATFET is disabled (in shipping mode) and indicated by setting BATFET\_DIS, one of the following events can enable BATFET to restore system power:

- 1. Plug in adapter
- 2. Clear BATFET\_DIS bit

3. Set REG\_RST bit to reset all registers including BATFET\_DIS bit to be default 0

4. A logic high to low transition on /QON pin with  $t_{\text{SHIPMODE}}$  deglitch time to enable BATFET to exit shipping mode

#### 10.8.3 BATFET Full System Reset

The BATFET functions as a load switch between battery and system when input source is not plugged–in. By changing the state of BATFET from on to off, systems connected to SYS can be effectively forced to have a poweron-reset. The /QON pin supports push-button interface to reset system power without host by changing the state of BATFET.

When the /QON pin is driven to logic low for  $t_{QON_RST}$  while input source is not plugged in and BATFET is enabled (BATFET\_DIS = 0), the BATFET is turned off for  $t_{BATFET_RST}$ and then it is re-enabled to reset system power. This function can be disabled by setting BATFET\_RST\_EN bit to 0.

### 10.8.4 /QON Pin Operations

The /QON pin incorporates two functions to control BATFET.

1. BATFET Enable: A /QON logic transition from high to low with longer than  $t_{SHIPMODE}$  deglitch turns on BATFET and exit shipping mode.

2. BATFET Reset: When /QON is driven to logic low by at least  $t_{QON_RST}$  while adapter is not plugged in (and BATFET\_DIS = 0), the BATFET is turned off for  $t_{BATFET_RST}$ . The BATFET is re-enabled after  $t_{BATFET_RST}$  duration. This

function allows systems connected to SYS to have poweron-reset. This function can be disabled by setting BATFET\_RST\_EN bit to 0.



Figure8 SC89890H /QON Timing

### 10.9 Power Good Indicator

The PG\_STAT bit goes HIGH to indicate a good input source when:

- VBUS above V<sub>VBUS\_UVLO</sub> below V<sub>VAC\_OVP</sub>
- VBUS above VBAT+VSLEEP (not in sleep)
- VBUS above V<sub>VBUSMIN</sub> (typical 3.8 V) when I<sub>BADSRC</sub> (typical 30 mA) current is applied (not a poor source)
- Completed input Source Type Detection

## 10.10 /INT

The SC89890H also has an alert mechanism that can output an interrupt signal via INT to notify the system of the operation by outputting a 256µs low-state INT pulse. All of the below events can trigger an INT output:

- USB/adapter source identified
- Good input source detected as described in power good indicator
- Input Removed
- Charge Complete
- VINDPM/IINDPM event detected(can be masked)
- Watchdog timer out, Safety timer out, OTG fault(VBUS overload, VBUS OVP, VBAT < VvBATLOW\_OTG), VBAT OVP, NTC COLD/HOT(Buck and Boost mode), Thermal shutdown, VBUS OVP, VBUS<VvBUSMIN</li>

When a fault occurs, the charger device sends out INT and keeps the fault state in REG until the host reads the fault register. The INT signal can be masked when the corresponding control bit is set. When a fault/status change occurs, the charger device sends out an INT pulse and keeps the state in REGOC until the host reads the registers. To read the current status, the host has to read REGOC two times consecutively. The first read reports the pre-existing

register status and the second read reports the current register status.

## **10.11 Protections**

### 10.11.1 Voltage and Current Monitoring in Buck Mode

### 10.11.1.1 Input Over voltage (ACOVP)

If VBUS voltage exceeds  $V_{VAC_OVP}$ , the device stops switching immediately. During input over voltage event (ACOV), the fault register CHRG\_FAULT bits are set to 01. An INT pulse is asserted to the host. The device will automatically resume normal operation once the input voltage drops back below the OVP threshold.

### 10.11.1.2 System Over voltage Protection(SYSOVP)

The charger device clamps the system voltage during load transient so that the components connect to system would not be damaged due to high voltage. SYSOVP threshold is 350 mV above minimum system regulation voltage when the system is regulate at V<sub>SYSMIN</sub> and above battery regulation voltage when battery charging is terminated. Upon SYSOVP, converter stops switching immediately to clamp the overshoot. The charger provides 30 mA discharge current to bring down the system voltage.

### 10.11.2 Voltage and Current Monitoring in Boost Mode

### 10.11.2.1 VBUS Soft Start

When the boost function is enabled, the device soft-starts boost mode to avoid inrush current.

### 10.11.2.2 VBUS Over Load Protection

The device monitors boost output voltage and other conditions to provide output short circuit and over voltage protection. The Boost build in accurate constant current regulation to allow OTG to adaptive to various types of load. If short circuit is detected on VBUS, the Boost begins hiccup mode until the short is removed.

### 10.11.2.3 VBUS Over Voltage Protection

When the VBUS voltage rises above regulation target and exceeds  $V_{OTG_OVP}$ , the device enters over voltage protection which stops switching, clears OTG\_CONFIG bit and exits boost mode. At Boost over voltage duration, the fault register bit (BOOST\_FAULT) is set high to indicate fault in boost operation. An INT is also asserted to the host.

#### 10.11.3 Thermal Regulation and Thermal Shutdown

#### 10.11.3.1 Thermal Protection in Buck Mode

The SC89890H monitors the internal junction temperature T<sub>J</sub> to avoid overheat the chip and limits the IC surface temperature in buck mode. When the internal junction temperature exceeds thermal regulation limit, the device lowers down the charge current. During thermal regulation, the actual charging current is usually below the programmed battery charging current. Therefore, termination is disabled, the safety timer runs at half the clock rate, and the status register THERM\_STAT bit goes high.

Additionally, the device has thermal shutdown to turn off the converter and BATFET when IC surface temperature exceeds  $T_{SD}$ . The fault register CHRG\_FAULT is set to 1 and an INT is asserted to the host. The BATFET and converter is enabled to recover when IC temperature is  $T_{SD_{-}HYS}$  below  $T_{SD}$ .

### 10.11.3.2 Thermal Protection in Boost Mode

The device monitors the internal junction temperature to provide thermal shutdown during boost mode. When IC junction temperature exceeds  $T_{SD}$ , the boost mode is disabled by setting OTG\_CONFIG bit low. When IC junction temperature is below  $T_{SD}$ - $T_{SD_HYS}$  the BATFET is enabled automatically to allow system to restore and the host can reenable OTG\_CONFIG bit to recover.

### 10.11.4 Battery Protection

### 10.11.4.1 Battery Over voltage Protection(VBATOVP)

The battery overvoltage limit is clamped at 4% above the battery regulation voltage. When battery over voltage occurs, the charger device immediately disables charging. The fault register BAT\_FAULT bit goes high and an INT is asserted to the host.

### 10.11.4.2 Battery Over discharge Protection

When battery is discharged below  $V_{BAT_DPL}$ , the BATFET is turned off to protect battery from over discharge. To recover from over-discharge latch-off, an input source plug-in is required at VBUS. The battery is charged with  $I_{BAT_SHORT}$ (typically 50mA) current when the VBAT <  $V_{BAT_SHORT}$ .

### 10.11.4.3 System Over current Protection

When the system is shorted or significantly overloaded (IBAT >  $I_{BAT_OCP}$  and  $I_{BAT_OCP_DGL}$  deglitch) so that the current

exceeds BATFET over current limit, the BATFET latches off. Section BATFET Enable (Exit Shipping Mode) can reset the latch-off condition and turn on BATFET.

## 10.12I2C Interface

### 10.12.1 I2C Interface

The IC features I2C interface, so the MCU or controller can control the IC flexibly. The 7-bit I2C address of the chip is 0x6A. The SDA and SCL pins are open drain and must be connected to the positive supply voltage via a current source or pull-up resistor. When the bus is free, both lines are HIGH. The I2C interface supports both standard mode (up to 100kbits) and fast mode (up to 400k bits with 5 k $\Omega$  pull up resistor at SCL pin and SDA pin respectively).

#### 10.12.2 Data Validity

The data on the SDA line must be stable during the HIGH period of the clock. The HIGH or LOW state of the data line can only change when the clock signal on the SCL line is LOW. One clock pulse is generated for each data bit transferred.



Figure 9 Bit transfer on the I2C bus

#### 10.12.3 START and STOP Conditions

All transactions begin with a START (S) and are terminated by a STOP (P). A HIGH to LOW transition on the SDA line while SCL is HIGH defines a START condition. A LOW to HIGH transition on the SDA line while SCL is HIGH defines a STOP condition.

START and STOP conditions are always generated by the master. The bus is considered to be busy after the START condition. The bus is considered to be free again a certain time after the STOP condition.



Figure 10 START and STOP conditions

#### 10.12.4 Byte Format

Every byte put on the SDA line must be eight bits long. The number of bytes that can be transmitted per transfer is unrestricted. Each byte must be followed by an Acknowledge bit. Data is transferred with the Most Significant Bit (MSB) first. If a slave cannot receive or transmit another complete byte of data until it has performed some other function, for example servicing an internal interrupt, it can hold the clock line SCL LOW to force the master into a wait state. Data transfer then continues when the slave is ready for another byte of data and releases clock line SCL.



Figure 11 Data transfer on the I2C bus

# 10.12.5 Acknowledge (ACK) and Not Acknowledge (NACK)

The acknowledge takes place after every byte. The acknowledge bit allows the receiver to signal the transmitter that the byte was successfully received and another byte may be sent. During data is transferred, the master can either be the transmitter or the receiver. No matter what it is, the master generates all clock pulses, including the acknowledge ninth clock pulse.

The transmitter releases the SDA line during the acknowledge clock pulse so the receiver can pull the SDA line LOW and it remains stable LOW during the HIGH period of this clock pulse.

When SDA remains HIGH during this ninth clock pulse, this is defined as the Not Acknowledge signal. The master can then generate either a STOP condition to abort the transfer, or a repeated START condition to start a new transfer.

#### 10.12.6 The slave address and R/W bit

Data transfers follow the format shown in below. After the START condition (S), a slave address is sent. This address is seven bits long followed by an eighth bit which is a data

direction bit (R/W) — a 'zero' indicates a transmission (WRITE), a 'one' indicates a request for data (READ). A data transfer is always terminated by a STOP condition (P) generated by the master. However, if a master still wishes to communicate on the bus, it can generate a repeated START condition (Sr) and address another slave without first generating a STOP condition.



Figure 12 complete data transfer



Figure 13 The first byte after the START procedure

#### 10.12.7 Single Read and Write

s	Slave Address (7 bits)	W(0)	ACK(0)	Reg Address (8 bits)	ACK(0)	Data (8 bits)	ACK(0)	Ρ
	from master to	o slave				from slave to mater		

#### Figure 14 Single Write



Figure 15 Single Read

If the register address is not defined, the charger IC send back NACK and go back to the idle state.

#### 10.12.8 Multi-Read and Multi-Write

The IC supports multi-read and multi-write for continuous registers.

s	Slave Address (7 bits)	W(0)	ACK(0)	Reg Address (8 bits)		ACK(0)		>			
	→ Dat:	a to Reg (8 bits)		ACK(0)	Data to Reg (8 bit		ACK(0)		Data to Reg Addr+N (8 bits)	ACK(0)	Р
from master to slave					1	from sla	ive to m	ater			

Figure 16 Multi-Write



Figure 17 Multi-Read

# 11 Register Map

Addr	Register	Туре	Default value @POR	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<u>00H</u>	INPUT CONTROL	R/W	01001000	EN_HIZ	EN_LIM	_LIM I <sub>IN_DPM</sub>					
<u>01H</u>	DPDM CONTROL1	R/W	000000X1		DP_DAC	•		DM_DAC		RESERVED	VINDPM_OS
<u>02H</u>	DPDM CONTROL2	R/W	00011X01	CONV_START	CONV_RATE	BOOST_FREQ	ICO_EN	HVDCP_EN	RESERVED	FORCE_DPDM	AUTO_DPDM_EN
<u>03H</u>	SYSTEM CONTROL1	R/W	00011010	FORCE_DSEL	WD_RST	OTG_CONFIG	CHG_CONFIG		V <sub>SYS_MIN</sub>		VVBATLOW_OTG
<u>04H</u>	ICC	R/W	00100010	EN_PUMPX			•	lcc			
<u>05H</u>	ITC & ITERM	R/W	00010100		ŀ	тс				Iterm	
<u>06H</u>	VBAT_REG	R/W	01011110			V <sub>BAT_REG</sub> V <sub>TC</sub>					V <sub>RECHG</sub>
<u>07H</u>	SYSTEM CONTROL2	R/W	10011101	EN_TERM	STAT_DIS	Т	Twd EN_TIMER T <sub>CHG</sub>			JEITA_ISET	
<u>08H</u>	TREG	R/W	00000011		BAT_COMP		VCLAMP			Т	REG
<u>09H</u>	SYSTEM CONTROL3	R/W	01000100	FORCE_ICCO	TMR2X_EN	BATFET_DIS	JEITA_VSET_ WARM	BATFET_DLY	BATFET_RST_E N	PUMPX_UP	PUMPX_DN
<u>0AH</u>	BOOST CONTROL	R/W	10110011		V <sub>BOOST</sub> PFN			PFM_OTG_DIS	IBOOST		
<u>0BH</u>	STAT1	R	XXXXXXXX		VBUS_STAT		CHF	RG_STAT	PG_STAT	RESERVED	VSYS_STAT
<u>0CH</u>	STAT2	R	xxxxxxx	WATCHDOG_ FAULT	OTG_FAULT	CHRG.	_FAULT	BAT_FAULT		NTC_FAULT	
<u>0DH</u>	VINDPM	R/W	00010010	FORCE_VIND PM				VINDPM			
<u>0EH</u>	STAT3	R	xxxxxxx	THERMAL_ST AT				ADC_VBA	Т		
<u>0FH</u>	STAT4	R	XXXXXXXX	RESERVED				ADC_VSY	S		
<u>10H</u>	STAT5	R	XXXXXXXX	RESERVED	RESERVED						
<u>11H</u>	STAT6	R	XXXXXXXX	VBUS_GD	BUS_GD ADC_VBUS						
<u>12H</u>	STAT7	R	XXXXXXXX	RESERVED				ADC_ICC			
<u>13H</u>	STAT8	R	XXXXXXXX	VINDPM_STAT	IINDPM_STAT				DPM_ICO		
<u>14H</u>	PN	R	0X100XXX	REG_RST	ICO_STAT		PART NUMBE	R	RESERVED	RESERVED	RESERVED

REG 00H( <u>Back to map</u> )	
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Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
7	R/W	EN_HIZ	0	Y	Y	Enable HI-Z Mode 0: Disable 1: Enable	Reset to default value when input source is plugged-in
6	R/W	EN_LIM	1	Y	Y	Enable ILIM Pin 0: Disable 1: Enable	
5	R/W		0	Y	Ν	1600mA	Input current limit
4	R/W		0	Y	Ν	800mA	Offset: 100 mA
3	R/W		1	Y	Ν	400mA	Range: 100 mA – 3.25A
2	R/W		0	Y	Ν	200mA	Default:500 mA
1	R/W		0	Y	Ν	100mA	,maximum input current limit, not typical.
0	R/W	IINDPM	0	Y	Ν	50mA	IINDPM bits are changed automatically after input source detection is completed Host can over-write IINDPM register bits after input source detection is completed.

# REG 01H(<u>Back to map</u>)

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes		
7	R/W		0	Y	N	000: HI-Z			
6	R/W		0	Y	N	001: 20K PULL DOWN			
						010: 0.6V			
		DP_DAC				011: 1.2V			
5	R/W	DI _DAG	0	Y	N	100: 2V			
5	r////		0	T	IN	101: 2.7V			
						110: 3.3V			
						111: RESERVED			
4	R/W		0	Y	Ν	000: HI-Z			
3	R/W	-	0	Y	Ν	001: 20K PULL DOWN			
								010: 0.6V	
						011: 1.2V			
2	R/W	DM_DAC	0	Y	N	100: 2V			
2	R/W		0	Ŷ	N	101: 2.7V			
						110: 3.3V			
						111: RESERVED			
1	R/W	RESERVED		Y	N				
						VINDPM Offset			
0	R/W	VINDPM_OS	1	Y	N	0: 400mV			
						1: 600mV			

# REG 02H(<u>Back to map</u>)

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
7	R/W	CONV_START	0	Y	Y	ADC Conversion Start Control 0: ADC conversion not active 1: Start ADC Conversion	This bit is read-only when CONV_RATE = 1. The bit stays high during ADC conversion and during input source detection.
6	R/W	CONV_RATE	0	Y	Y	ADC Conversion Rate Selection 0: One shot ADC conversion 1: Start 1s Continuous Conversion	
5	R/W	BOOST_FREQ	0	Y	Y	Boost Mode Frequency Selection 0: 1.5MHz 1: 500KHz	
4	R/W	ICO_EN	1	Y	Ν	Input Current Optimizer (ICO) Enable 0: Disable ICO Algorithm 1: Enable ICO Algorithm	
3	R/W	HVDCP_EN	1	Y	Ν	High Voltage DCP handshake when DCP is identified 0: Disable HVDCP handshake 1: Enable HVDCP handshake	
2	R/W	RESERVED					
1	RW1C	FORCE_DPDM	0	Y	Y	Force DP/DM Detection 0: Not in DP/DM detection 1: Force DP/DM detection	
0	R/W	AUTO_DPDPM_ EN	1	Y	N	0: Disable DP/DM detection when VBUS is plugged-in 1: Enable DP/DM detection when VBUS is plugged-in	

# REG 03H(<u>Back to map</u>)

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
7	R/W	FORCE_DSEL	0	Y	N	<ul><li>0: Allow DSEL pin output to drive low</li><li>1: Force DSEL pin output to</li></ul>	
						drive high	
						I2C Watchdog Timer Reset	
6	RW1C	WD_RST	0	Y	Y	0: Normal	
						1: Reset	
						Boost Mode Configuration	
5	R/W	OTG_CONFIG	0	Y	Y	0: OTG Disable	
						1: OTG Enable	
						Charge Enable Configuration	
4	R/W	CHG_CONFIG	1	Y	Y	0: Charge Disable	
						1: Charge Enable	
3	R/W	-	1	Y	Ν	000:2.6V	
2	R/W	-	0	Y	Ν	001:2.8V	
						010:3V	
		V <sub>SYS_MIN</sub>				011:3.2V	
1	R/W	V SYS_MIN	1	Y	N	100:3.4V	
I	12/10			ľ	IN IN	101:3.5V	
						110:3.6V	
						111:3.7V	
	5.44					Minimum Battery Voltage (falling) to exit boost mode	
0	R/W	V <sub>VBATLOW_OTG</sub>	0	Y	Y	0: 2.8V	
						1: 2.5V	

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
7	R/W	EN_PUMPX	0	Y	Y	0: Disable Current pulse control 1: Enable Current pulse control (PUMPX_UP and PUMPX_DN)	
6	R/W		0	Y	Y	3840 mA	
5	R/W		1	Y	Y	1920 mA	
4	R/W		0	Y	Y	960 mA	Fast Charge Current
3	R/W	I <sub>cc</sub>	0	Y	Y	480 mA	Offset: 0mA
2	R/W		0	Y	Y	240 mA	Range: 0mA – 5040mA
1	R/W		1	Y	Y	120 mA	
0	R/W		0	Y	Y	60 mA	

# REG 04H(<u>Back to map</u>)

# REG 05H(Back to map)

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
7	R/W		0	Y	Y	480 mA	
6	R/W		0	Y	Y	240 mA	Trickle Current
5	R/W	I <sub>TC</sub>	0	Y	Y	120 mA	Offset: 60mA
4	R/W		1	Y	Y	60 mA	Range: 60mA – 960mA
3	R/W		0	Y	Y	480 mA	
2	R/W	I <sub>TERM</sub>	1	Y	Y	240 mA	Termination Current
1	R/W		0	Y	Y	120 mA	Offset: 30mA
0	R/W	-	0	Y	Y	60 mA	Range: 30mA – 930mA

# REG 06H(<u>Back to map</u>)

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
7	R/W		0	Y	Y	512 mV	
6	R/W		1	Y	Y	256 mV	
5	R/W		0	Y	Y	128 mV	Charge Voltage
4	R/W	V <sub>BAT_REG</sub>	1	Y	Y	64 mV	Offset: 3.840V
3	R/W	-	1	Y	Y	32 mV	Range: 3.84V-4.848V
2	R/W	-	1	Y	Y	16 mV	-
1	R/W	V <sub>TC</sub>	1	Y	Y	Battery TC to CC Charge Threshold 0: 2.8V 1: 3.0V	
0	R/W	V <sub>RECHG</sub>	0	Y	Y	Battery Recharge Threshold (below Charge Voltage) 0: 100mV 1: 200mV	

# REG 07H(<u>Back to map</u>)

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
7	R/W	EN_TERM	1	Y	Y	Charging Termination Enable 0: Disable 1: Enable	
6	R/W	STAT_DIS	0	Y	Y	0: Enable STAT pin function 1: Disable STAT pin function	
5	R/W		0	Y	Y	I2C Watchdog Timer Setting	
4	R/W	T <sub>WD</sub>	1	Y	Y	00: Disable watchdog timer 01: 40s 10: 80s 11: 160s	
3	R/W	EN_TIMER	1	Y	Y	Charging Safety Timer Enable 0: Disable 1: Enable	
2	R/W		1	Y	Y	Fast Charge Timer Setting	
1	R/W	Т <sub>СНБ</sub>	0	Y	Y	00: 5 hrs 01: 8 hrs 10: 12 hrs 11: 20 hrs	
0	R/W	JEITA_ISET	1	Y	Y	JEITA Cool Temperature Current Setting 0: 50% of ICC 1: 20% of ICC	

# REG 08H(<u>Back to map</u>)

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
7	R/W		0	Y	Y	80 mΩ	IR Compensation Resistor
6	R/W	BAT_COMP	0	Y	Y	40 mΩ	Setting Range: 0-140mΩ
5	R/W		0	Y	Y	20 mΩ	Default: 0mΩ
4	R/W		0	Y	Y	128 mV	IR Compensation Resistor
3	R/W	VCLAMP	0	Y	Y	64 mV	Setting
2	R/W		0	Y	Y	32 mV	Range: 0-224mV Default: 0mV
1	R/W		1	Y	Y	Thermal Regulation	
						00: 60°C	
	DAA	T <sub>REG</sub>		Y	X	01: 80°C	
0	R/W		1	Ý	Y	10: 100°C	
						11: 120°C	

# Reg 09H(<u>Back to map</u>)

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
-	DWAO	50005 100	0	X	X	0: Do not force ICO	
7	RW1C	FORCE_ICO	0	Y	Y	1: Force ICO	
						Safety Timer Setting during DPM or Thermal Regulation	
6	R/W	TMR2X_EN	1	Y	Y	0: Safety timer not slowed by 2X during input DPM or thermal regulation or JEITA	
						1: Safety timer slowed by 2X during input DPM or thermal regulation or JEITA	
5	R/W	BATFET_DIS	0	Y	Ν	0: Allow BATFET turn on 1: Turn off BATFET	
						JEITA Warm Temperature Voltage Setting	
4	R/W	JEITA_VSET_WA RM	0	Y	Y	0: Set Charge Voltage to VREG-200mV during JEITA warm temperature	
						1: Set Charge Voltage to VREG during JEITA high temperature	
3	R/W	BATFET_DLY	0	Y	Ν	0: Turn off BATFET immediately when BATFET_DIS bit is set	
5			0			1: Turn off BATFET after t <sub>BATFET_DLY</sub> (typ. 10 s) when BATFET_DIS bit is set	
2	R/W	BATFET_RST_E	1	Y	N	0: Disable BATFET reset function	
2		N				1: Enable BATFET reset function	
	DW40			N N	X	Current pulse control voltage up enable	
1	RW1C	PUMPX_UP	0	Y	Y	0: Disable	
						1: Enable	
0	DW40		<u> </u>	N N	X	Current pulse control voltage down enable	
0	RW1C	PUMX_DN	0	Y	Y	0: Disable	
						1: Enable	

# Reg 0AH(<u>Back to map</u>)

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
7	R/W		1	Y	Y	800 mV	
6	R/W		0	Y	Y	400 mV	Offset:3.9V
5	R/W	V <sub>BOOST</sub>	1	Y	Y	200 mV	Range:3.9V-5.4V
4	R/W		1	Y	Y	100 mV	
3	R/W	PFM_OTG_DIS	0	Y	N	PFM mode allowed in boost mode 0: Allow PFM in boost mode 1: Disable PFM in boost mode	
2	R/W		0	Y	N	000: 0.5A	
1	R/W		1	Y	N	001: 0.75A	
0	R/W	I <sub>BOOST</sub>	1	Y	Ν	010: 1.2A 011: 1.4A 100: 1.65A 101: 1.875A 110: 2.15A 111: 2.45A	

# Reg 0BH(<u>Back to map</u>)

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
7	R		Х	NA	NA	VBUS Status register	
6	R		Х	NA	NA	000: No Input	
5	R	VBUS STAT	X	NA	NA	001: USB Host SDP 010: USB CDP (1.5A) 011: USB DCP (3.25A) 100: HVDCP 101: Unknown Adapter (500mA) 110: Non-Standard Adapter (1A/2A/2.1A/2.4A) 111: OTG Note: Software current limit is reported in IINLIM register	
4	R		Х	NA	NA	Charging status:	
3	R	CHRG_STAT	x	NA	NA	00: Not Charging 01:TC Charging 10:CC Charging 11: Charge Termination	
2	R	PG_STAT	x	NA	NA	Power Good status: 0: Power Not Good 1:Power Good	
1	R	RESERVED					
0	R	VSYS_STAT	x	NA	NA	0:Not in V <sub>SYS_MIN</sub> regulation (BAT > V <sub>SYS_MIN</sub> ) 1:in V <sub>SYS_MIN</sub> regulation (BAT < V <sub>SYS_MIN</sub> )	

# Reg 0CH(<u>Back to map</u>)

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
7	R	WATCHDOG_FA	х	NA	NA	0: Normal	
		ULT	~			1: Watchdog timer expiration	
						0: Normal	
6	R	BOOST_FAULT	х	NA	NA	1: VBUS overloaded in OTG, or VBUS OVP, or battery is too low (any conditions that we cannot start boost function)	
5	R		х	NA	NA	00: Normal;	
		CHRG_FAULT		NA		01: input fault (VAC OVP or VBAT < VBUS < 3.8 V);	
4	R		Х		NA	10: Thermal shutdown;	
						11: Charge Safety Timer Expiration	
0	<b>D</b>		v	NA	NA	0: Normal	
3	R	BAT_FAULT	Х	NA		1: BATOVP	
2	R		х	NA	NA	NTC Fault Status	
1	R		х	NA	NA	Buck Mode:	
						000: Normal	
						010: NTC Warm	
						011: NTC Cool	
		NTC_FAULT				101: NTC Cold	
0	R		х	NA	NA	110: NTC Hot	
						Boost Mode:	
						000: Normal	
						101: NTC Cold	
						110: NTC Hot	

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
7	R/W	FORCE_VINDPM	0	Y	Ν	VINDPM Threshold Setting Method 0: Run Relative VINDPM Threshold 1: Run Absolute VINDPM Threshold	Reset to default value when input source is plugged-in
6	R/W		0	Y	N	6400 mV	Absolute VINDPM Threshold
5	R/W		0	Y	N	3200 mV	
4	R/W		1	Y	N	1600 mV	Offset: 2.6V
3	R/W	VINDPM	0	Y	N	800 mV	Effective Range: 3.9V-
2	R/W		0	Y	N	400 mV	15.3V
1	R/W		1	Y	N	200 mV	Reset to default value when input source is
0	R/W		0	Y	Ν	100 mV	plugged-in

# Reg 0DH(<u>Back to map</u>)

# Reg 0EH(<u>Back to map</u>)

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
7	R	THERMAL_STAT	х	NA	NA	Thermal Regulation Status 0: Normal	
		_				1: In Thermal Regulation	
6	R		Х	NA	NA	1280 mV	
5	R		Х	NA	NA	640 mV	_
4	R		Х	NA	NA	320 mV	Offset: 2.304V
3	R	ADC_VBAT	х	NA	NA	160 mV	Range: 2.304V
2	R		Х	NA	NA	80 mV	(0000000) – 4.848V (111111)
1	R		Х	NA	NA	40 mV	
0	R		х	NA	NA	20 mV	

# Reg 0FH(<u>Back to map</u>)

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
7	R	RESERVED	х	NA	NA		
6	R		х	NA	NA	1280 mV	
5	R		х	NA	NA	640 mV	
4	R		Х	NA	NA	320 mV	Offset: 2.304V
3	R	ADC_VSYS	Х	NA	NA	160 mV	Range: 2.304V
2	R		Х	NA	NA	80 mV	(0000000) – 4.848V (111111)
1	R		Х	NA	NA	40 mV	
0	R		х	NA	NA	20 mV	

# Reg 10H(<u>Back to map</u>)

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
7	R	RESERVED	х	NA	NA		
6	R	RESERVED	х	NA	NA		
5	R	RESERVED	х	NA	NA		
4	R	RESERVED	х	NA	NA		
3	R	RESERVED	х	NA	NA		
2	R	RESERVED	х	NA	NA		
1	R	RESERVED	х	NA	NA		
0	R	RESERVED	х	NA	NA		

# Reg 11H(<u>Back to map</u>)

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
7	R	VBUS_GD	х	NA	NA	VBUS Good Status 0: Not VBUS attached 1: VBUS Attached	
6	R		х	NA	NA	6400 mV	
5	R		х	NA	NA	3200 mV	
4	R		x	NA	NA	1600 mV	ADC VBUS
3	R	ADC_VBUS	х	NA	NA	800 mV	Offset: 2.6V
2	R		х	NA	NA	400 mV	Range 2.6V (0000000) – 15.3V (1111111)
1	R		х	NA	NA	200 mV	
0	R		х	NA	NA	100 mV	

# Reg 12H(<u>Back to map</u>)

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
7	R	RESERVED					
6	R		х	NA	NA	3200 mA	
5	R		x	NA	NA	1600 mA	ADC conversion of
4	R		х	NA	NA	800 mA	Charge Current (IBAT)
3	R	ADC_ICC	х	NA	NA	400 mA	Offset: 0mA
2	R		х	NA	NA	200 mA	Range 0mA (0000000) – 6350mA (1111111)
1	R		х	NA	NA	100 mA	
0	R		х	NA	NA	50 mA	

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
7	R	VINDPM_STAT	Х	NA	NA	VINDPM Status 0: Not in VINDPM 1: VINDPM	
6	R	IINDPM_STAT	Х	NA	NA	IINDPM Status 0: Not in IINDPM 1: IINDPM	
5	R		х	NA	NA	1600 mA	Input Current Limit in
4	R		х	NA	NA	800 mA	effect while Input Current Optimizer
3	R		х	NA	NA	400 mA	(ICO) is enabled
2	R	IDPM_ICO	х	NA	NA	200 mA	Offset: 100mA (default) Range 100mA (0000000) – 3.25A (1111111)
1	R		х	NA	NA	100 mA	
0	R		х	NA	NA	50 mA	

# Reg 13H(<u>Back to map</u>)

# Reg 14H(<u>Back to map</u>)

Bit	Mode	Bit Name	Default value @POR	Reset by REG_RST	Reset by WTD	Description	Notes
7	RW1C	REG RST	0	NA	NA	Register Reset 0: Keep current register setting 1: Reset to default register value and reset safety timer	
6	R	ICO_STAT	x	NA	NA	Input Current Optimizer (ICO) Status 0: ICO is in progress 1: ICO Finished	
5	R		1	NA	NA	Part Number	
4	R	PN	0	NA	NA		
3	R		0	NA	NA		
2	R	RESERVED					
1	R	RESERVED					
0	R	RESERVED					

# **12 MECHANICAL DATA**





### **RECOMMENDED FOOTPRINT(Unit mm)**